### LOW-POWER ADCS FOR MASSIVE MIMO SYSTEMS JAN CRANINCKX



## MASSIVE HARDWARE NEEDS FOR MASSIVE MIMO



# ENERGY-EFFICIENT WIRELESS ADCS

- Performance requirements
  - >10 ENOB
  - I 00's MS/s
  - Low power consumption (mW)
- Architecture choice
  - SAR: successive approximation
  - N comparisons needed for an N-bit ADC
  - Nanoscale CMOS offers enough speed to reach required sampling rates

## OUTLINE

#### • ADC Architecture

- Channel implementation
- Calibration methods
- Implementations and measurements
- Conclusions

## START FROM SAR ADC FOR EFFICIENCY



### PIPELINE FOR SPEED



## **INTERLEAVE FOR MORE SPEED**



## ADC CHANNEL (I): COARSE SAR SAMPLES AND GENERATES RESIDUE



- 6 cycle binary scaled SAR
- Ib redundancy in backend
  - robust to comparator errors

## SINGLE-ENDED DAC SWITCHING SCHEME FOR LOW ENERGY, GOOD SPEED



## DAC LINEARITY IS CRITICAL

- Total capacitance = 3.5 pF
  - Approx. 13b noise
- Calibration for 3 MSBs
  - Programmable capacitance
  - Digital coefficients binary scaled
  - LSBs sufficiently matched



## CONTROLLER REPLACED BY COMPARATORS + DELAYS

- Each comparator activated at specific common-mode
  - Calibrate offset at this common-mode
- Reduces power consumption



## ADC CHANNEL (2): DYNAMIC RESIDUE AMPLIFICATION



- Amplify and sample residue
- Small output swing  $\rightarrow$  relaxed linearity
- $\circ$  Uncertain gain  $\rightarrow$  match coarse SAR LSB to fine SAR MSB in calibration
- Fast amplification with low noise at low power

## DYNAMIC AMPLIFIER = INTEGRATION DURING CERTAIN TIME



$$E = V_{dd} \cdot I_{D} \cdot t_{int}$$

Low noise requires high  $g_m$  or long integration time

## **INVERTER IS GOOD TRANSCONDUCTOR**



- PMOS and NMOS contribute gm
- Shared bias current  $\rightarrow$  efficiency

# DIFFERENTIALLY, COMMON MODE FEEDBACK, START/END INTEGRATION



### ADC CHANNEL (3): FINE SAR QUANTIZES RESIDUE



- 8 cycle binary SAR w/ intermediate noise
  - 400 μV r.m.s.
- 2 cycle binary SAR w/ low noise
  - $\circ~$  200  $\mu V$  r.m.s.
- Ib redundancy between
- Similar implementation to coarse SAR

## POWER EFFICIENCY ENABLED BY CALIBRATION



- Calibration corrects
  - DAC mismatch
  - Comparator thresholds: 6 coarse stage, 7 fine stage
  - Residue amplifier gain
  - Channel gain

## COMPARATOR OFFSET CALIBRATION OFF-LINE

- Short ADC input
- Use DAC to generate correct common-mode
- Change comparator thresholds to average Dout = 0.5
- In second stage: Also cancels amplifier offset
- $\circ$  Sensitive to PVT



## INTER-STAGE GAIN CALIBRATION OFF-LINE



- Short ADC input
- Use coarse DAC to generate +0.5/-0.5 coarse LSB amplifier input
- Change gain to obtain DoutI-Dout2 = I fine MSB
- Sensitive to PVT

## MSB CAPACITANCE CALIBRATION OFF-LINE



- Short ADC input
- MSB comparator noise  $\rightarrow$  2 residue values
- Use second stage to measure MSB S(LSBs)
- Change MSB size to obtain DoutI-Dout2 = I fine MSB
- Insensitive to PVT

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# 1.7mW 11b 250MS/s in 40nm





## 2. ImW IIb 410MS/s in 28nm WITH BACKGROUND ON-CHIP CALIBRATION ENGINE





## 2.3mW 14b 200 MS/s in 28nm







- SAR architecture and its improvements enable power-efficient data conversion for wireless systems
  - Dynamic and asynchronous controller
  - Pipelined
  - Time-interleaved
- Nanoscale digital CMOS for SoC integration
- $\circ$  I to 2mW power consumption
  - Negligible in the overall system power



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- B.Verbruggen et al., "A 2.1 mW 11b 410 MS/s dynamic pipelined SAR ADC with background calibration in 28nm digital CMOS," 2013 Symposium on VLSI Circuits, Kyoto, 2013, pp. C268-C269.
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